Systems Engineering Grounding and Wires

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### Reference Node

• Consider imperfect conductor: 10 m $\Omega$ /cm

$$0V \text{ rail} \begin{array}{c|c} & & & \\ & & I_1 (20\text{mA}) \end{array} \begin{array}{c} & I_2 (10\text{mA}) \end{array} \begin{array}{c} & I_3 (10\text{mA}) \end{array} \begin{array}{c} \text{power supply} \\ \text{connection} \end{array}$$

Resulting voltages

$$V_C = (i_1 + i_2 + i_3)10 \text{ m}\Omega = 400 \,\mu\text{V}$$
$$V_B = 400 \,\mu\text{V} + (i_1 + i_2)10 \,\text{m}\Omega = 700 \,\mu\text{V}$$
$$V_A = 700 \,\mu\text{V} + (i_1)10 \,\text{m}\Omega = 900 \,\mu\text{V}$$

# Reference Node

Is this a problem?

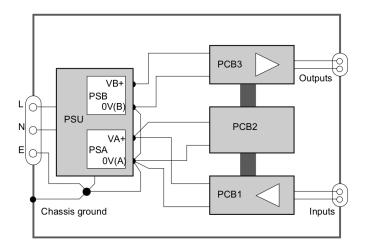
- Amps instead of milli- or microamps
- Resistance in ohms instead of milliohms
- Depending on the SNR!

The order of blocks may be important!

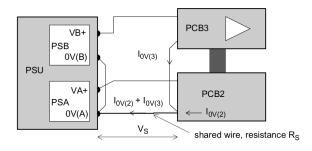
# Grounding inside one unit

PCB1 Input signal conditioning

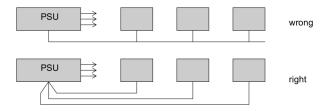
- PCB2 Microcontroller board
- PCB3 High-current output drivers



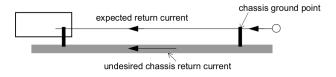
### Power return



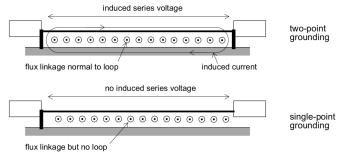
#### Star vs bus returns



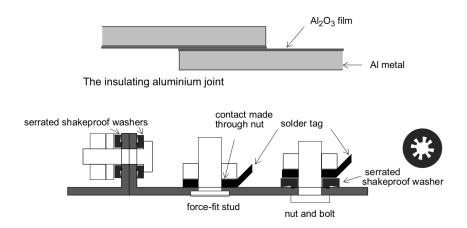
# Single-Point Chassis Ground



- Impedances depend on frequency
- Joints in chassis affecte by corrosion
- ► Ground loop:

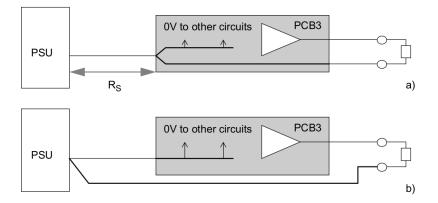


# Aluminium and aluminium joints



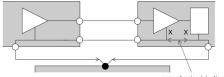
# **Output Signal Ground**

#### Avoid common impedances



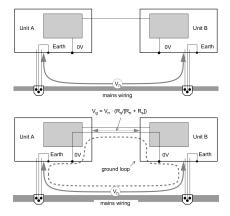
# Inter-Board Signals

- ► First possibility: do nothing. Ground returns through PS
- Insert inter-board ground connection
  - High-speed digital communications
  - Precision analogue signals
  - Drawbacks
  - There is alternative path for power return
  - Ground loops
  - Solutions
  - Separate returns
  - Differential connections / optocouplers / 1:1 trafo



ground noise injection here

# Ground Between Units



- ► Float (!)
- Differential link
- Galvanic isolation
  - transformer
  - optocouplers
  - fibre optic link

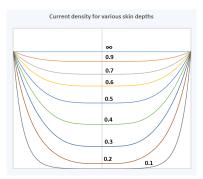
### Cables

Туре	Dia (mm)	I Nom	l Melt	R	L	f max (kHz)
AWG 32	0.2	0.33	$\sim 5$	0.54	1.8	430
AWG 26	0.4	1.2	$\sim 20$	0.11	1.7	107

Meaning of maximum frequency?

# Skin Effect

- AC current tends to concentrate near the surface of the conductor
- Skin depth  $\delta$ : equivalent thickness
- Equivalent tubular conductor at DC





# Skin Effect/ 2

$$\delta = \sqrt{\frac{\rho}{\pi\mu f}} = \frac{6.10 \times 10^{-2}}{\sqrt{f[\text{Hz}]}}$$

▶ At 10 MHz,  $\delta = 0.02$  mm

Equivalent area for AWG32 (diameter 0.2 mm)

$$A_{eq} = \pi imes 0.2^2 - \pi imes (0.2 - 0.02)^2 = \pi imes 0.0076$$

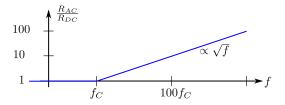
Area at DC

$$A_{f=0} = \pi \times 0.2^2 = \pi \times 0.04$$

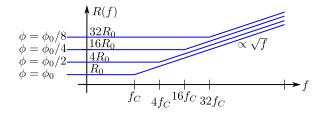
The resistance at 10 MHz is  $\frac{0.04}{0.076} = 5.3$  times higher than at DC.

### Frequency Response

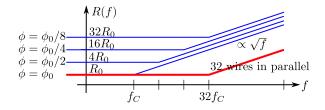
- AC current tends to concentrate near the surface of the conductor
- Skin depth  $\delta$ : equivalent thickness
- Equivalent tubular conductor at DC



One thick or several thin cables /1



# One thick or several thin cables /1



- Litz Wire
- From the german: Litzendraht. Expensive. Weaving pattern
- Used for coils in the MHz range: Required values of L mean significant wire length
- AM radio
- Induction coils